

11/07/11

second order Butterworth low pass filter

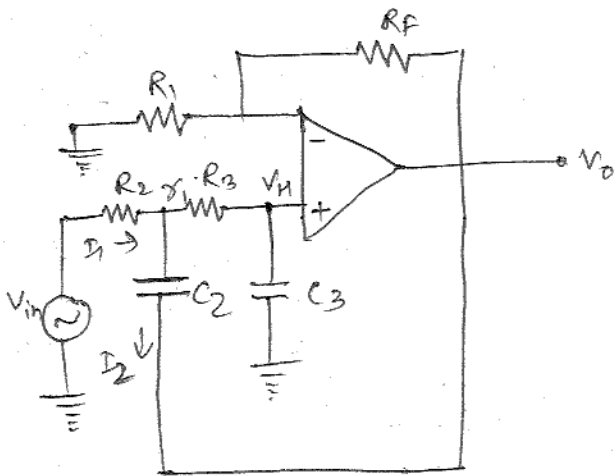


Fig ①

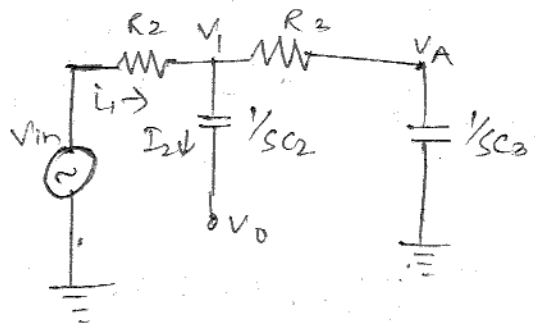


Fig ②

using KCL in fig ②

$$I_1 = I_2 + I_3$$

$$\frac{V_{in} - V_1}{R_2} = \frac{V_1 - V_0}{1/sC_2} + \frac{V_1 - V_A}{R_3} \quad \text{--- ①}$$

use potential rules at pt. A

$$V_A = V_1 \left[ \frac{1/sC_3}{R_3 + 1/sC_3} \right]$$

$$V_A = \frac{V_1}{sR_3C_3 + 1} \quad \text{--- ②}$$

$$V_1 = V_A (sR_3C_3 + 1) \quad \text{--- ③}$$

sub. eqn ③ in eqn ①

$$\frac{V_{in} - V_A (sR_3C_3 + 1)}{R_2} = \frac{V_A (sR_3C_3 + 1) - V_0}{1/sC_2} + \frac{V_A (1 + C_3R_3s) - V_A}{R_3}$$

$$\frac{V_{in}}{R_2} - \frac{V_A (sR_3C_3 + 1)}{R_2} = sC_2 \left( V_A (R_3C_3S + 1) - V_0 \right) + \frac{V_A (1 + R_3C_3S - 1)}{R_3}$$

$$\frac{V_{in}}{R_2} + V_0 sC_2 = V_A \left[ \frac{1 + sR_3C_3}{R_2} + sC_2 (1 + sR_3C_3) + \frac{1 + R_3C_3S}{R_3} - \frac{1}{R_3} \right]$$

$$\frac{V_{in}}{R_2} + V_0 sC_2 = V_A \left[ \frac{R_3 (1 + sR_3C_3) + sC_2 R_2 R_3 (1 + R_3C_3S) + R_2 (1 + R_3C_3S) - R_2}{R_2 R_3} \right]$$

$$R_3 V_{in} + V_0 R_2 R_3 sC_2 = V_A \left[ (1 + R_3C_3S) (R_3 + sC_2 R_2 R_3 + R_2) - R_2 \right]$$

$$V_A = \frac{R_3 V_{in} + V_0 R_2 R_3 sC_2}{\left[ (1 + R_3C_3S) (R_3 + R_2 R_3 sC_2 + R_2) - R_2 \right]}$$

In general:

$$V_0 = A_F \cdot V_A$$

$$V_0 = A_F \frac{R_3 V_{in} + V_0 R_2 R_3 sC_2}{\left[ (1 + R_3C_3S) (R_3 + R_2 R_3 sC_2 + R_2) - R_2 \right]}$$

$$A_F R_3 V_{in} = V_0 \left[ 1 - \frac{A_F R_3 R_2 sC_2}{(1 + R_3C_3S) (R_3 + R_2 R_3 sC_2 + R_2) - R_2} \right]$$

$$A_F R_3 V_{in} = V_0 \left[ (1 + R_3C_3S) (R_3 + R_2 R_3 sC_2 + R_2) - R_2 - A_F R_3 R_2 sC_2 \right]$$

$$\frac{V_0}{V_{in}} = \frac{A_F R_3}{\left[ (1 + sR_3C_3) (R_3 + R_2 R_3 sC_2 + R_2) - R_2 - A_F R_3 R_2 sC_2 \right]}$$

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$$\frac{V_o}{V_{in}} = \frac{A_F R_3}{R_3 + sR_3^2 C_3 + R_3 R_2 C_2 s + sR_3^2 C_3 C_2 R_2 + R_2 + sR_3 R_2 C_3 - R_2 - sR_2 R_3 C_2 A_F}$$

$$= \frac{A_F}{1 + sR_3 C_3 + R_2 C_2 s + sR_3 C_3 C_2 R_2 + sR_3^2 C_3 - A_F sR_2 C_2}$$

$$\frac{V_o}{V_{in}} = \frac{A_F}{s^2 + s(R_2 C_3 + R_2 C_2 + R_2 C_3 - A_F R_2 C_2) + \frac{1}{R_3 C_3 C_2 R_2}} \quad \text{--- (A)}$$

In general.

$$\frac{V_o(s)}{V_i(s)} = \frac{A}{s^2 + 2\zeta \omega_n s + \omega_n^2} \quad \text{--- (B)}$$

Comparing A and B.

$$\omega_n^2 = \frac{1}{R_2 R_3 C_2 C_3}$$

$$\omega_n = \omega_c$$

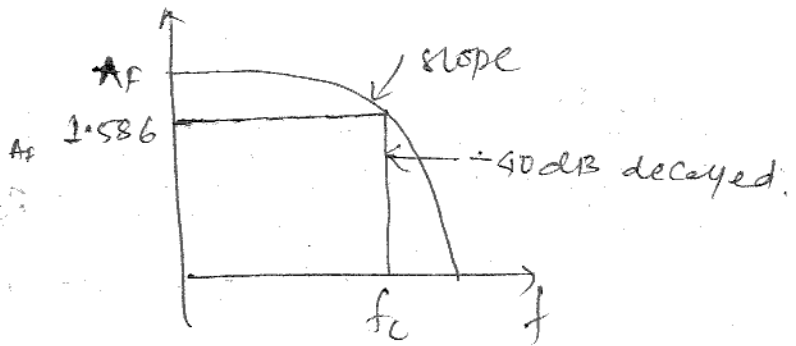
$$\omega_c^2 = \frac{1}{R_2 R_3 C_2 C_3}$$

$$f_c^2 = \frac{1}{(2\pi)^2 R_2 R_3 C_2 C_3}$$

$$f_c = \frac{1}{\sqrt{(2\pi)^2 R^2 C^2}}$$

$$\left[ \begin{array}{l} \because R_2 = R_3 = R \\ C_2 = C_3 = C \end{array} \right]$$

$$\therefore \boxed{f_c = \frac{1}{2\pi RC}}$$



$$0.4 \epsilon^2 \omega_n^2 = \left[ \frac{s(RC + RC + RC - AF \cdot RC)}{R^2 C^2} \right]^2$$

$$4 \epsilon^2 \frac{1}{R^2 C^2} = \frac{s(3RC - AF)}{R^2 C^2}$$

$$\epsilon = \frac{\sqrt{s(3RC - AF)}}{2}$$

$$2 \epsilon \omega_n = \frac{\cancel{s 3RC} + \cancel{s 3RC} + \cancel{s 3RC} - s \cdot 3RC(3 - AF)}{R^2 C^2}$$

$$2 \epsilon \omega_n = 0(3 - AF)$$

$$4 \epsilon \frac{1}{RC} = \frac{0(3 - AF)}{RC}$$

$$\epsilon = \frac{0(3 - AF)}{2}$$

$$\frac{1}{\sqrt{2}} = \frac{2 - A_f}{2}$$

$$\boxed{A_f = 1.586}$$

$$= 20 \log(1.586)$$

$$= -40 \text{ dB / decade}$$

$$A_f = 1 + \frac{R_f}{R_1}$$

$$1.586 = 1 + \frac{R_f}{R_1}$$

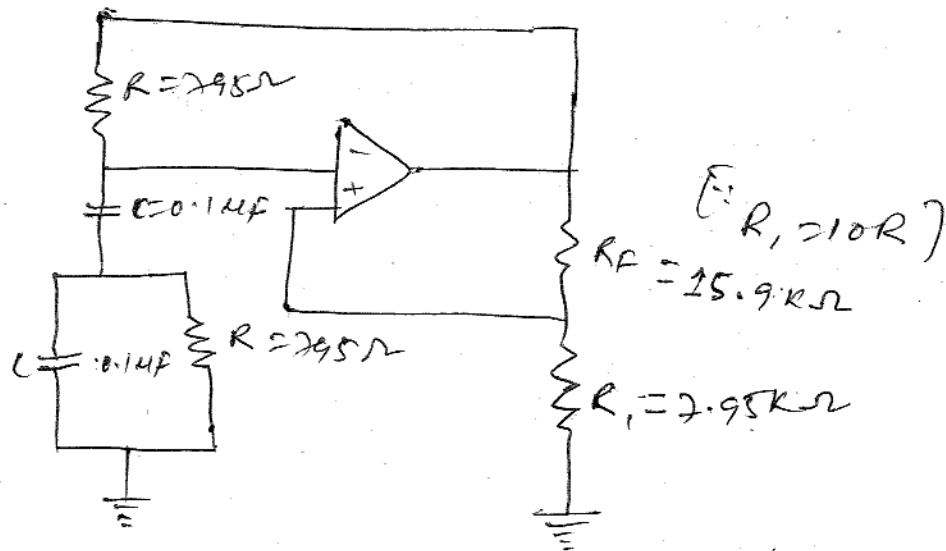
$$\therefore \boxed{R_f = 0.586 R_1}$$

19/07/11.

Procedure Design procedure for second order low pass filter (LPF).

- 1) choose the cut-off frequency.
- 2) select  $R_2 = R_3 = R_1$ , and  $C_2 = C_3 = C$
- 3) select  $A_f = 1.586$ ,  $R_f = 0.586 R_1$ .
- 4) select  $R_1 \leq 1 \text{ M}\Omega$

Q1] use  $\pm 15V$  power supply and IC 741.  
 Design Wein Bridge oscillator for 2kHz freq.  
 Assume  $C = 0.1\mu F$ .



$$R_f = 2R_1$$

$$f = 2\text{kHz}$$

$$C = 0.1\mu F$$

$$f = \frac{1}{2\pi RC}$$

$$R = \frac{1}{2\pi f C} = 795\Omega$$

$$R = 795\Omega$$

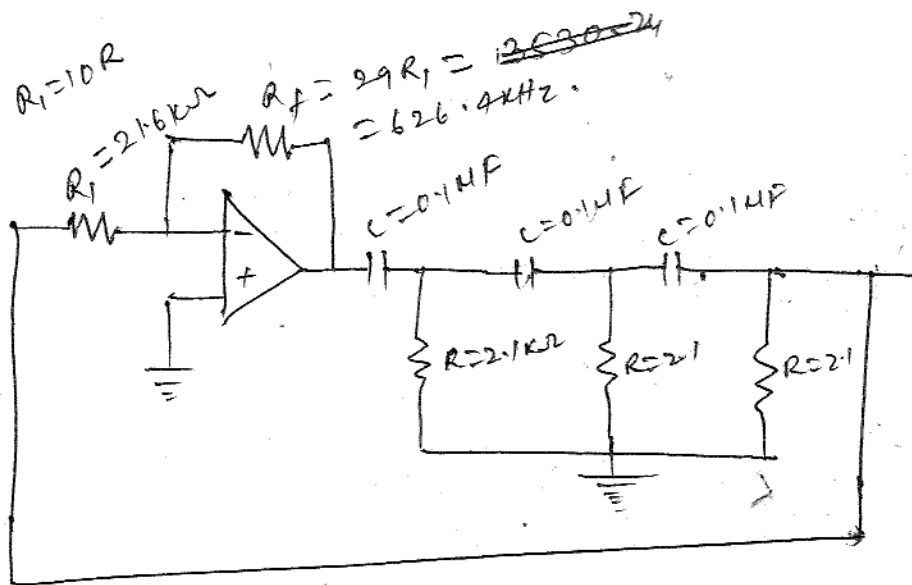
Q2] Design a phase shift oscillator,  
 oscillates at 300Hz and  $C = 0.1\mu F$

$$f = \frac{1}{2\pi\sqrt{6}RC}$$

$$R = \frac{1 \times 10^6}{2 \times 3.14 \times \sqrt{6} \times 300 \times 0.1}$$

$$R = \frac{10^6}{461.48} = 2166.9\Omega$$

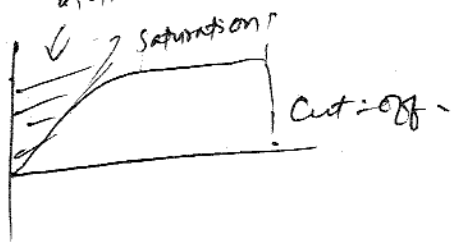
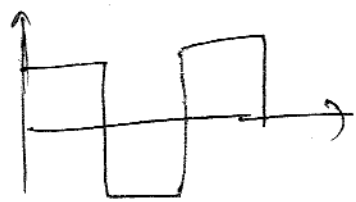
$$R = 2.167\text{k}\Omega$$



AC phase shift oscillator

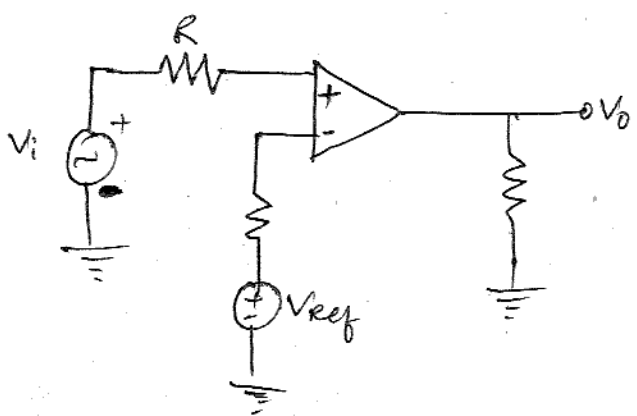
COMPARATOR

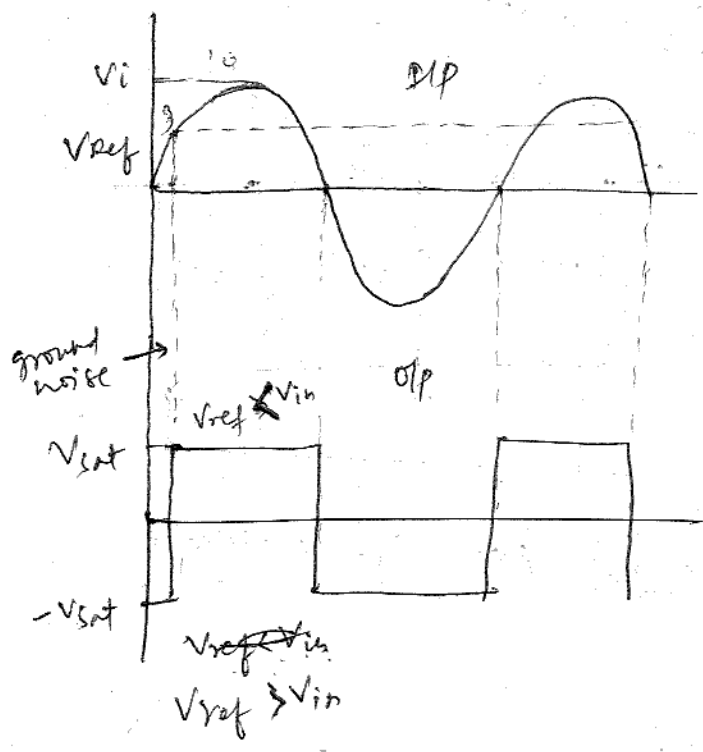
Inverting      Non-inverting  
 active (as an amplifier)



Non-linear  
if  $V_{ref} > V_{in}$

Non-Inverting Comparator





Inverting.