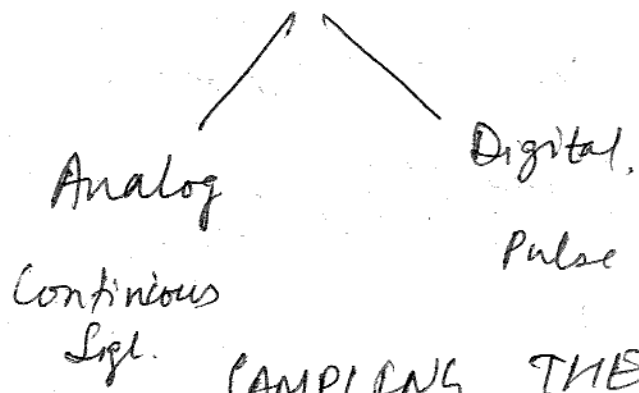


# Communication



## SAMPLING THEOREM! -

A Band limited signal of finite energy has no frequency component higher than  $\omega$  Hertz, is completely described by specifying the value of the signal at instant of Time  $\frac{1}{2\omega}$  seconds.

$\omega \rightarrow$  max<sup>m</sup> frequency.

$$T_s = \frac{1}{2\omega}$$

$$f_s = 2\omega.$$

Sampling Theory can also be stated as band limited signal of finite energy which has more frequency component than  $\omega$  Hertz, may be completely recovered from the knowledge of  $1f_s$  samples taken at the rate of  $2\omega$  samples/sec.

A Continuous time signal can be completely represented in its samples & recovered back if the sampling frequency is  $f_s \geq 2w$ .

Here,

$f_s$  is the sampling frequency, &

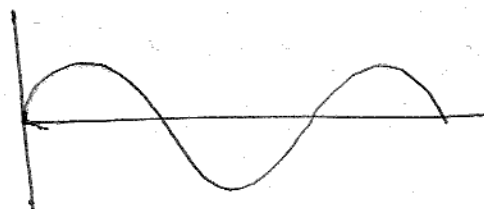
$w$  is the max<sup>m</sup> frequency.

### Proof Sampling theorem

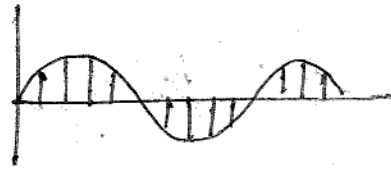
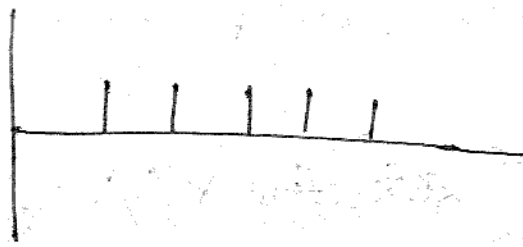
Let  $x(t)$  be a continuous time signal which doesn't contain any frequency component higher than  $w$  hertz.

A sampling function samples the signal at regularly at the rate of  $f_s$  samples / sec. i.e.

$$T_s = \frac{1}{f_s} \Rightarrow T_s = \text{Sampling period}$$



sampling i/p



The pulse train of Pulse is rep as,

$$f(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT_s) \quad \text{--- (1)}$$

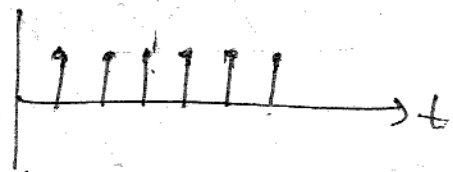
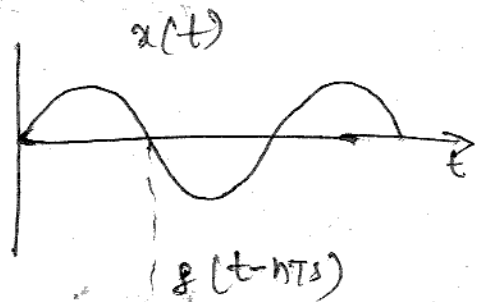
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The impulse train pulse is

$$f(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT_s) \quad \text{--- (1)}$$

The waveform (3) represented as

$$x_s(t) = \delta(t - nT_s) x(nT_s)$$

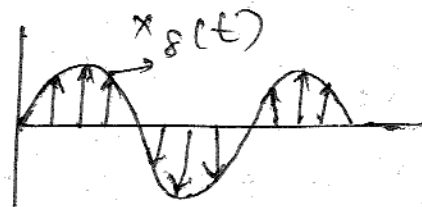


The Fourier transform of

the pulse train pulse is:-

$$X(f) = f_s \sum_{n=-\infty}^{\infty} \delta(t - nT_s)$$

(where,  $f_s = \frac{1}{T_s}$ )



# Tabby Fourier transform of waveform (3)

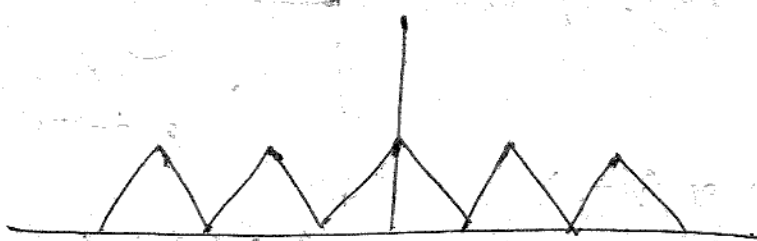
$$X_s(f) = f_s \sum_{n=-\infty}^{\infty} x(f - nf_s) \quad \text{--- (2)}$$

This Eq shows that the process of uniformly sampling a continuous time signal results in a periodic spectrum with periods equal to the sampling rate.

The Fourier of signal  $x(t)$  results in

$$X(f - nf_s) = X(f), \text{ where}$$

$$f = 0, \pm f_s, \pm 2f_s$$



Thus the same spectrum  $X(f)$  appears as,

$$f = 0, f = \pm f_s, f = \pm 2f_s$$

Eq (2) can be written as,

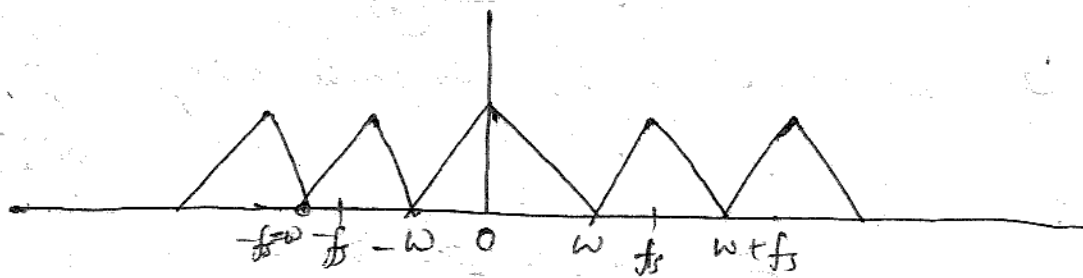
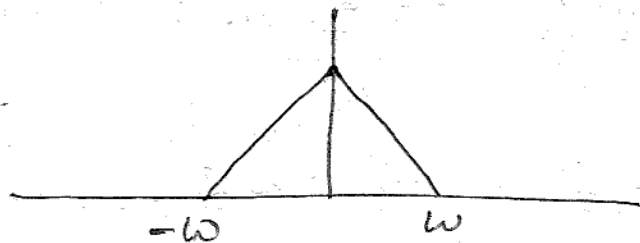
$$X_s(f) = f_s X(f) + f_s X(f \pm f_s) + f_s X(f \pm 2f_s) + \dots \quad \text{--- (3)}$$

Eq<sup>n</sup> (2) can also be written as

$$X_s(f) = f_s X(f) + \sum_{\substack{n=-\infty \\ n \neq 0}}^{\infty} f_s X(f - n f_s) \quad \text{--- (3)}$$

w.k.t

Sampling frequency,  $f_s = 2W$      $T_s = \frac{1}{2W}$



⇒ The  $f_s$  represent spectrum of sample frequency.

By the def<sup>n</sup> of Fourier transform of C.T s/g.

$$FT[X(t)] = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt$$

The Fourier transform of D.T signal is

$$F[X_s(t)] = \sum_{-\infty}^{\infty} x(nT_s) e^{-j2\pi f n T_s}$$

From Eq<sup>n</sup> (3), we can write,

$$f_s X(f) = X_s(f) - \sum_{n=0}^{\infty} f_s X(f_s - n f_s)$$

$$X(f) = \frac{1}{f_s} X_s(f) - \sum_{n=0}^{\infty} X(f_s - n f_s)$$

Substitute,  
 $f_s = 2\omega$

$$X(f) = \frac{1}{2\omega} X_s(f) - \sum_{n=0}^{\infty} X(f - n f_s)$$

$$X(f) = \frac{1}{2\omega} X_s(f) \quad \text{for } -\omega \leq f \leq \omega$$

Eq<sup>n</sup> (4) can be written as, with a help of

$$X(f) = \frac{1}{2\omega} \sum_{n=-\infty}^{\infty} x(nT_s) e^{-j2\pi f n T_s}$$

$$T_s = \frac{1}{f_s}, \quad \ln T_s = \frac{1}{2\omega}$$

$$X(f) = \frac{1}{2\omega} \sum_{n=-\infty}^{\infty} x\left(\frac{n}{2\omega}\right) e^{-j2\pi f n \frac{1}{2\omega}}$$

$$X(f) = \frac{1}{2\omega} \sum_{n=-\infty}^{\infty} x\left(\frac{n}{2\omega}\right) e^{-\frac{j\pi f n}{\omega}}$$